



Tunable all-optical microwave logic gates based on nonreciprocal topologically protected edge modes

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Abstract: All-optical logic gates have been studied intensively owing to their potential to enable broadband, low-loss and high-speed communications. However, poor tunability has remained a key challenge in this field. In this work, we propose a Y-shaped structure composed of Yttrium Iron Garnet (YIG) layers that can serve as tunable all-optical logic gates, including, but not limited to, OR, AND and NOT gates, by applying external magnetic fields to magnetize the YIG layers. Our findings reveal that these logic gates are founded on protected one-way edge modes, where by tuning the wavenumber k of the operating mode to a sufficiently small (or even zero) value, the gates can become nearly immune to nonlocal effects. This not only enhances their reliability but also allows for maintaining extremely high precision in their operations. Furthermore, the operating band itself of the logic gates is also shown to be tunable. We introduce a straightforward and practical method for controlling and switching these gates between "work", "skip", and "stop" modes. These findings have potentially significant implications for the design of high-performance and robust all-optical microwave communication systems.

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1. Introduction

Since the invention of the transistor in 1947, there has been an unprecedented upsurge in electronic communications based on electrical signals [1,2]. However, with the development of integrated circuits, transistors are becoming increasingly miniaturized, resulting in increased energy waste. Furthermore, electronic communications still suffer from defects such as cross-talk [3]. All-optical communications would benefit from advantages, such as high-speed signal processing, error-free transmission [4], parallel computation [5], and low loss [6], making them a potential candidate for next-generation communication technology.

All-optical logic gates (LGs) are an important component of integrated optical circuits, and have received considerable attention in recent years, with interesting results in this field. Researchers have constructed various types of all-optical LGs, such as photonic crystal and Mach-Zehnder interferometer structures, using nonlinear processes [7–9] and/or interferometry [10–13], and

have implemented all basic logic operations. However, most LGs suffer from low contrast ratios (CRs), typically less than 30 dB. This is understandable because reflections are unavoidable in conventional optical LGs, and imperfections in their manufacturing affect the accuracy of the gates to some extent, particularly in nonlinearity-based LGs [14–16]. In many studies on sub-wavelength all-optical LGs, researchers often neglect the impact of nonlocal effects on logical operations. While this is generally true in near-wavelength cases, non-local effects should be considered when the device's scale is subwavelength or even deep-subwavelength. In fact, the impact of nonlocal effects on nonreciprocal/one-way surface magnetoplasmons (SMPs) has been widely discussed in the past several years [17–19]. SMPs are edge modes sustained in magneto-optical (MO) heterostructures, and many interesting and meaningful results, such as slow light [20–22] and rainbow trapping [23,24], have been discovered. Recently, we proposed a method to implement (sub-wavelength) all-optical logic operations using one-way SMP modes [25]. This type of one-way electromagnetic (EM) mode has been proven to be topologically protected [26,27] in the microwave regime by several research groups, and no significant impact of non-local effects has been observed. Therefore, in this paper, we focus on such nonlocality-immune SMPs to study tunable LGs. Additionally, since guided wave modes have only one transmission direction, the problem of preparation process defects is well overcome, and unidirectional modes are immune to backscattering. More importantly, all-optical LGs based on unidirectional EM modes theoretically have an infinite contrast ratio, which means unparalleled accuracy.

Note that in Ref. [25], the designed all-optical LGs relied on Yttrium Iron Garnet (YIG) with remanence. Consequently, although unidirectional SMPs-based all-optical LGs were implemented using MO heterostructures, their lack of tunability hindered their application in future integrated optical circuits. In this paper, we propose a Y-shaped structure composed of three YIG layers under different bias magnetic fields and theoretically analyze the dispersion relation in the three arms, which are all YIG-YIG heterostructures. We observe interesting phenomena, such as reverse propagation direction, and close and/or reopen one-way regions. More importantly, we discover highly tunable characteristics of the Y-shaped structure and the LGs, which are confirmed by full-wave simulation. Our proposed tunable subwavelength LGs have the potential to be utilized in the realization of high-performance and programmable all-optical microwave communication systems, offering enhanced flexibility and functionality.

We note here that a previous work [25] focused on a distinct issue, namely the examination of the feasibility of unidirectional mode-based LGs, whereas here we explore the possibility of realizing *tunable* LGs. Furthermore, in our present investigation (particularly evident in Fig. 3 and Eqs. (2)–(8)) we show that randomly (arbitrarily) selecting and testing external magnetic fields to implement LGs is a laborious and unreliable approach. Instead, we show that the external magnetic fields/asymptotic frequencies in the Y-shaped structure must meet specific conditions in order to achieve logic operations, necessitating rigorous theoretical derivation and verification. Guided by such an analysis, we introduce the novel concept of switchable states/modes in the LGs, which is important for enhancing the efficiency and *programmability* of such gates.

2. Physical model and topologically protected SMPs

The Y-shaped configuration is a commonly used physical model in the field of all-optical LGs, which has been extensively studied in recent decades [11,28–31]. In Fig. 1(a), we propose a Y-shaped YIG-based model that enables tunable all-optical logic operations. The model comprises three straight arms, each containing two layers of YIG. Unlike our previous work [25], where YIG with remanence was used, all the YIG layers in this study are subjected to an external magnetic field (H_0) to further enhance the tunability of the LGs. It should be noted that metals can always be considered as perfect electric conductor (PEC) walls in the microwave regime [32]. For simplicity, as shown in Fig. 1(b), the arm with YIG layers having the same magnetization is

referred to as 'EYYE-s', where "E" represents the PEC boundary, "Y" represents YIG, and "s" symbolizes the same magnetization direction. Similarly, the structure with YIG layers having opposite magnetization directions is labeled 'EYYE-r'. To achieve basic logic operations based on one-way modes, the key is to establish two separate one-way channels that allow efficient transfer of the EM wave/signal. The question then arises as to how to design suitable arms and how to efficiently tune the structure according to our needs.

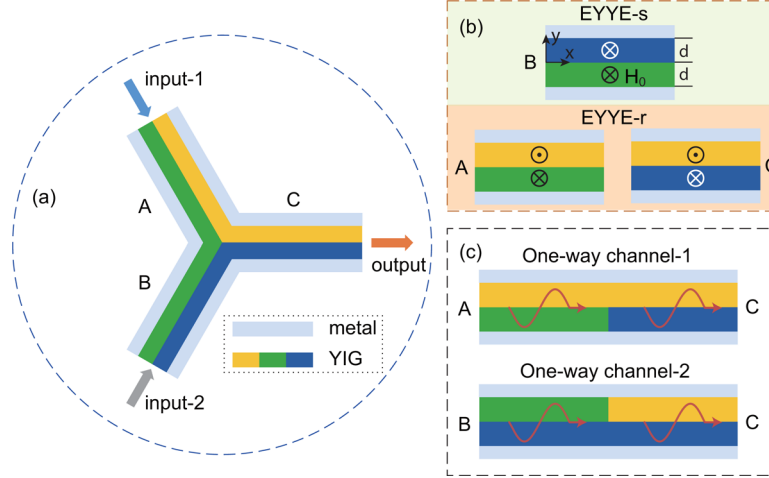


Fig. 1. (a) The schematic of the Y-shaped structure of all-optical logic operations. (b) Two types of arms are shown, i.e. the 'EYYE-s' and the 'EYYE-r'. (c) Pre-designed two one-way channels in our proposed structure. Note that, in this paper, we use ω_0^a , ω_0^b , and ω_0^c to clarify the precession angular frequencies (ω_0) for green-colored YIG, yellow-colored YIG and blue-colored YIG layers, respectively.

To achieve this, one must first study the dispersion relation of the SMPs in those arms. The 'EYYE-r' contains two layers of YIG with two different relative permeability and for the lower ($\bar{\mu}_a$) and upper ($\bar{\mu}_b$) YIG, we have

$$\bar{\mu}_a = \begin{bmatrix} \mu_1^a & -i\mu_2^a & 0 \\ i\mu_2^a & \mu_1^a & 0 \\ 0 & 0 & 1 \end{bmatrix}, \quad \bar{\mu}_b = \begin{bmatrix} \mu_1^b & i\mu_2^b & 0 \\ -i\mu_2^b & \mu_1^b & 0 \\ 0 & 0 & 1 \end{bmatrix} \quad (1)$$

where $\mu_1 = 1 + \frac{\omega_m(\omega_0 - i\nu\omega)}{(\omega_0 - i\nu\omega)^2 - \omega^2}$ and $\mu_2 = \frac{\omega_m\omega}{(\omega_0 - i\nu\omega)^2 - \omega^2}$. ω , $\omega_m = \mu_0\gamma M_s$ ($M_s = 1780$ gauss, the saturation magnetization of YIG [32,33]), ν and $\omega_0 = \mu_0\gamma H_0$ refer respectively to the angular frequency, the characteristic circular frequency, the damping factor, and the precession angular frequency [32]. Please note that the superscripts 'a' and 'b' represent the lower and upper layers, respectively. In this paper, we assume that the magnetic-field direction in the lower layer is permanently oriented in the -z direction. By applying Maxwell's equations and three boundary conditions in the 'EYYE-r' arm, one can easily calculate the dispersion relation of the SMPs sustained on the YIG-YIG interface. The dispersion relation takes the following form

$$\mu_v^a \left[\frac{\mu_2^b}{\mu_1^b} k + \frac{\alpha_b}{\tanh(\alpha_b d)} \right] + \mu_v^b \left[\frac{\mu_2^a}{\mu_1^a} k + \frac{\alpha_a}{\tanh(\alpha_a d)} \right] = 0 \quad (2)$$

where $\alpha_a = \sqrt{k^2 - \varepsilon_m \mu_v^a k_0^2}$, $\alpha_b = \sqrt{k^2 - \varepsilon_m \mu_v^b k_0^2}$, and $\mu_v = \mu_1 - \mu_1^2/\mu_2$. Equation (2) reveals that the SMPs in the 'EYYE-r' arm exhibit different propagation properties for opposite wavenumbers,

i.e., $k_1 = -k_2$, which is a well-known nonreciprocity effect. More importantly, adjusting the external magnetic field can create a special one-way region where the waves propagate in only one specific direction. The asymptotic frequencies (AFs) of the SMPs in the 'EYYE-r' arm can be derived and calculated from Eq. (2). We found four AFs, which can be described by the following equations:

$$\omega_{\text{sp}}^{(+)} = \begin{cases} \omega_{\text{sp}}^{(+1)} = \omega_0^{\text{a}} + \omega_{\text{m}} \\ \omega_{\text{sp}}^{(+2)} = \omega_0^{\text{b}} + \omega_{\text{m}} \end{cases} \quad (3)$$

$$\omega_{\text{sp}}^{(-)} = \begin{cases} \omega_{\text{sp}}^{(-1)} = \frac{(\omega_0^{\text{a}} + \omega_0^{\text{b}} + \omega_{\text{m}}) + \sqrt{(\omega_0^{\text{a}} + \omega_0^{\text{b}} + \omega_{\text{m}})^2 - 2(2\omega_0^{\text{a}}\omega_0^{\text{b}} + \omega_0^{\text{a}}\omega_{\text{m}} + \omega_0^{\text{b}}\omega_{\text{m}})}}{2} \\ \omega_{\text{sp}}^{(-2)} = \frac{(\omega_0^{\text{a}} + \omega_0^{\text{b}} + \omega_{\text{m}}) - \sqrt{(\omega_0^{\text{a}} + \omega_0^{\text{b}} + \omega_{\text{m}})^2 - 2(2\omega_0^{\text{a}}\omega_0^{\text{b}} + \omega_0^{\text{a}}\omega_{\text{m}} + \omega_0^{\text{b}}\omega_{\text{m}})}}{2} \end{cases} \quad (4)$$

ω_{sp}^{+} and ω_{sp}^{-} indicate the AF as $k \rightarrow +\infty$ and $k \rightarrow -\infty$, respectively. In fact, the value of ω_{sp}^{+} corresponds to the zero point of μ_v^{a} or μ_v^{b} . Similarly, the dispersion relation of the SMPs in the 'EYYE-s' arm can be directly obtained from Eq. (2) by replacing μ_2^{b} , μ_1^{b} , μ_v^{b} , and α_{b} with $-\mu_2^{\text{c}}$, μ_1^{c} , μ_v^{c} , and α_{c} , respectively. In this case, the permeability ($\bar{\mu}_{\text{c}}$) of the upper YIG has the same form as $\bar{\mu}_{\text{a}}$, and the corresponding dispersion equation can be written as follows:

$$\mu_v^{\text{a}} \left[\frac{-\mu_2^{\text{c}}}{\mu_1^{\text{c}}} k + \frac{\alpha_{\text{c}}}{\tanh(\alpha_{\text{c}} d)} \right] + \mu_v^{\text{c}} \left[\frac{\mu_2^{\text{a}}}{\mu_1^{\text{a}}} k + \frac{\alpha_{\text{a}}}{\tanh(\alpha_{\text{a}} d)} \right] = 0 \quad (5)$$

There are also four potential AFs in the 'EYYE-s' arm, which have the following form:

$$\omega_{\text{sp}}^{(+)} = \begin{cases} \omega_{\text{sp}}^{(+1)} = \omega_0^{\text{a}} + \omega_{\text{m}} \\ \omega_{\text{sp}}^{(+2)} = \frac{(\omega_0^{\text{c}} - \omega_0^{\text{a}}) + \sqrt{(\omega_0^{\text{c}} - \omega_0^{\text{a}})^2 + 2(2\omega_0^{\text{a}}\omega_0^{\text{c}} + \omega_0^{\text{a}}\omega_{\text{m}} + \omega_0^{\text{c}}\omega_{\text{m}})}}{2} \end{cases} \quad (6)$$

$$\omega_{\text{sp}}^{(-)} = \begin{cases} \omega_{\text{sp}}^{(-1)} = \omega_0^{\text{c}} + \omega_{\text{m}} \\ \omega_{\text{sp}}^{(-2)} = \frac{(\omega_0^{\text{a}} - \omega_0^{\text{c}}) + \sqrt{(\omega_0^{\text{a}} - \omega_0^{\text{c}})^2 + 2(2\omega_0^{\text{a}}\omega_0^{\text{c}} + \omega_0^{\text{a}}\omega_{\text{m}} + \omega_0^{\text{c}}\omega_{\text{m}})}}{2} \end{cases} \quad (7)$$

Based on Eqs. (4) and (5), we plot the dispersion curves for the SMPs in both the 'EYYE-r' and 'EYYE-s' arms as $d = 0.02\lambda_{\text{m}}$ ($\lambda_{\text{m}} = 2\pi c/\omega_{\text{m}} = 60$ mm) and $\nu = 0$ (lossless condition). It is worth noting that the thickness of YIG in the z direction can be chosen arbitrarily (e.g. at subwavelength scale), and the results presented in this paper can be easily extended to three dimensions by setting the terminal material to be a metal or PEC [26,34,35]. Three different values of ω_0 (H_0) are applied in the three arms, and for convenience, we introduce a simple notation - '[α, β, θ]' - in which α and β represent the absolute values of the normalized ω_0 ($\bar{\omega}_0 = \omega_0/\omega_{\text{m}}$) for the lower and upper YIG, while θ could be either 'r' referring to the 'EYYE-r' arm or 's' referring to the 'EYYE-s' arm. For example, [0.6, 0.3, r] in Fig. 2(a) implies that the dispersion curve is calculated in the 'EYYE-r' arm, where $\omega_0^{\text{a}} = 0.6\omega_{\text{m}}$ and $\omega_0^{\text{b}} = 0.3\omega_{\text{m}}$. In Fig. 2(a), the green and blue stars represent $\omega_{\text{sp}}^{(+)}$ and $\omega_{\text{sp}}^{(-)}$, respectively. The red and black lines indicate the dispersion curves of SMPs on the YIG-YIG interface, and due to the deep-subwavelength thickness of the YIG layers, the bulk zones are significantly compressed [23,36]. Therefore, it is believed that almost all the SMPs on the red and black lines are one-way EM modes except for the SMPs located near the resonant frequencies of YIG ($\omega_{\text{s}} = \sqrt{\omega_0^2 + \omega_0\omega_{\text{m}}}$), which are marked by black arrows. As depicted in the inset of Fig. 2(a), the case of [0.6, 0.3, r] can be treated as one of the input arms (arm 'A') of the Y-shaped heterostructure. We also calculate the dispersion curves for the other arms in Figs. 2(b) and 2(c). As a result, similar to the first case, there are two one-way regions in both cases. However, in the case of [0.6, 0.4, s], the EM waves within the lower one-way region have negative group velocities ($v_{\text{g}} < 0$).

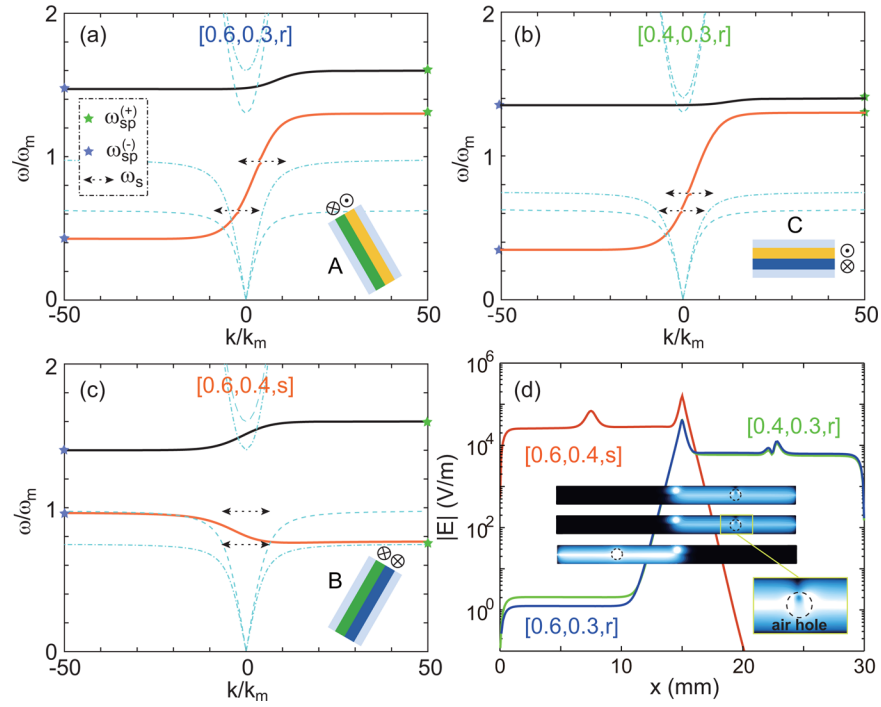


Fig. 2. (a-c) The dispersion diagrams of three arms are shown, in which the lower YIG has ω_0 values of $0.6\omega_m$, $0.4\omega_m$, and $0.6\omega_m$, respectively, while the upper YIG has ω_0 values of $0.3\omega_m$, $0.3\omega_m$, and $0.4\omega_m$, respectively. Note that 'r' and 's' indicate the 'EYYE-r' arm and 'EYYE-s' arm, respectively, and the magnetization orientation of the lower YIG is permanently -z. The cyan lines represent the edge of the bulk zones, while the black arrows indicate the location of $\omega = \omega_s$. Stars show the corresponding asymptotic frequencies in each case. (d) The simulated electric field distributions of the three cases are shown for $f = 0.8f_m$ ($f_m = 5$ GHz). The other parameters are (a-c) $d = 0.02\lambda_m$, $\nu = 0$ and (d) $\nu = 0.001\omega_m$.

Based on Eqs. (3), (4), (6), and (7) as mentioned earlier, the one-way regions are defined by the AFs (green and blue stars in Fig. 2). For the three cases discussed above, the regions are: (a) $[0.428f_m, 1.3f_m]$ and $[1.472f_m, 1.6f_m]$, (b) $[0.3475f_m, 1.3f_m]$ and $[1.3525f_m, 1.4f_m]$, and (c) $[0.766f_m, 0.966f_m]$ and $[1.4f_m, 1.6f_m]$. Therefore, to design two one-way channels, the frequencies used must be located within the $[0.766f_m, 0.966f_m]$ region (the red line region in Fig. 2(c)). In addition, the loss effect and the robustness of the one-way propagation of SMPs are examined using full-wave simulations, as illustrated in Fig. 2(d). In this case, we consider $\nu = 0.001\omega_m$ and $f = 0.8f_m$, and air holes with a radius of $r = 0.5$ mm ($\sim 0.008\lambda_m$) are placed on the YIG-YIG interface. The simulation results show good agreement with the theoretical analysis, and the imperfections have a negligible impact on the one-way SMPs. It is also worth noting that recent studies have questioned the robustness of one-way SMPs, with the nonlocal effect being a major focus of these works [17,18]. Here, we emphasize that the one-way SMPs studied in this paper are theoretically topologically protected, which has been theoretically demonstrated [37,38] and experimentally proved [26,39] by many groups. This nonlocality-immune property is particularly evident in cases where the waveguide is relatively thick or the wavenumber (k) is relatively small [19]. Our proposed LGs in this paper are believed to be largely unaffected by nonlocal effects, given the tunability of the SMPs, which will be discussed in the next subsection.

3. Tunability of the SMPs

In our theoretical analysis, we have shown that a Y-shaped structure consisting of magnetized YIG layers can support two independent one-way channels, making it suitable as a logical gate [25]. More importantly, benefiting from the tunability of the topologically protected one-way SMPs, the proposed LGs should be easily tunable by changing the bias magnetic fields. In the following sections, we demonstrate the tunability of our proposed LGs in detail. Firstly, we study the impact of H_0 on AFs, which always define the one-way regions. As displayed in Fig. 3(a), four AFs in the 'EYYE-r' arm 'A' ('C'), are plotted as a function of ω_0^a (ω_0^c) and ω_0^b . Figure 3(b) depicts the similar relationship between AFs and ω_a and ω_c . To differentiate between the four distinct AFs in Eqs. (3,4) ('EYYE-r') and (6,7) ('EYYE-s'), we use the names $\omega_{sp}^{(+1)}$, $\omega_{sp}^{(+2)}$, $\omega_{sp}^{(-1)}$, and $\omega_{sp}^{(-2)}$. Notably, as ω_0 (H_0) changes, the values of AFs and their numerical relationships may also change. This can lead to a reversal of the group velocity and the transmission direction of EM signals in LGs. Therefore, any changes in the AFs can affect the functionality of the LGs.

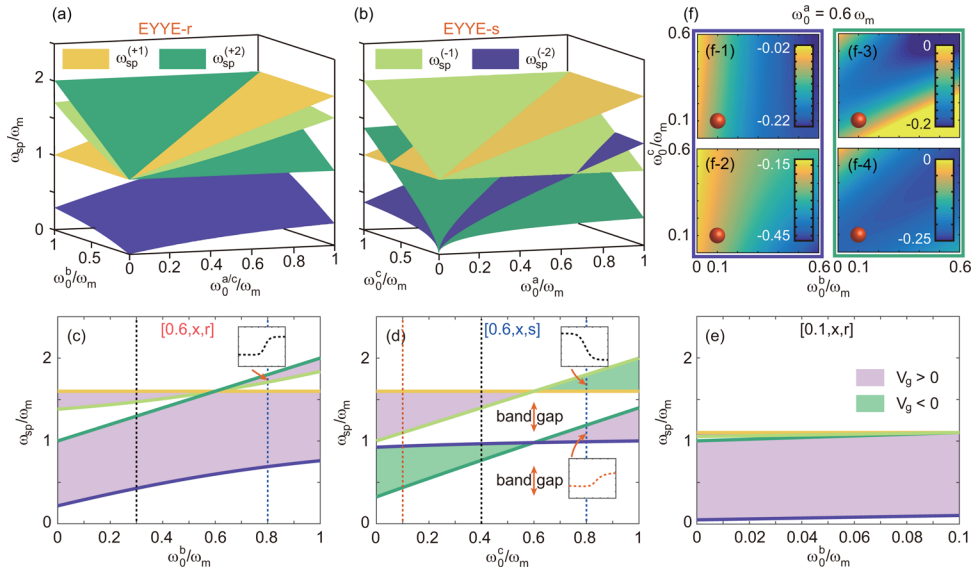


Fig. 3. (a,b) The asymptotic frequencies (AFs) are plotted as a function of ω_0^a (or ω_0^c) and ω_0^b for (a) the 'EYYE-r' arm and (b) the 'EYYE-s' arm. (c-e) AFs are plotted as a function of ω_0^b when (c,d) $\omega_0^a = 0.6\omega_m$ and (e) $\omega_0^c = 0.1\omega_m$. (f) Four constructed equations (y_1 , y_2 , y_3 , and y_4) as shown in Eq. (8) are plotted as functions of ω_0^b and ω_0^c when $\omega_0^a = 0.6\omega_m$.

To illustrate the changes in AFs and one-way regions, we set the lower YIG ω_0 to $0.6\omega_m$ and assume $0 < \omega_0 < \omega_m$ for the upper YIG in both 'EYYE-r' (Fig. 3(c)) and 'EYYE-s' (Fig. 3(d)) arms. As ω_0 (ω_0^b) of the upper YIG varies from 0 to $0.6\omega_m$, the lower one-way region gradually widens, while the upper one-way region becomes smaller and eventually closes at $\omega_0^b = 0.6\omega_m$. The black dashed line represents the [0.6, 0.3, r] case discussed earlier in Fig. 2(a), in which two clear one-way regions are present (excluding the local area near $\omega = \omega_s$). As ω_0^b increases further, for the 'EYYE-r' arm, the first one-way region is compressed slightly, while a new one-way region bounded by $\omega_{sp}^{(-1)}$ and $\omega_{sp}^{(+2)}$ emerges with a forward propagation direction ($v_g > 0$). The inset of Fig. 3(c) displays a zoomed-in dispersion curve for the case of $\omega_0^b = 0.8\omega_m$ (blue line). In contrast, the 'EYYE-s' arm behaves differently. As shown in Fig. 3(d), when ω_0^c (in the upper YIG) is increased, the propagation direction of SMPs in the lower one-way region changes from backward ($v_g < 0$) to forward ($v_g > 0$), and the one-way region closes and reopens.

Similar phenomena of reversing propagation direction and close-reopen one-way regions are observed in the higher regime as well. The black and blue dashed lines in Fig. 3(d) indicate cases where $\omega_0^c = 0.4\omega_m$ and $\omega_0^c = 0.8\omega_m$, respectively, with $\omega_0^a = 0.6\omega_m$ in both cases. The insets in Fig. 3(d) show the reversed one-way regions and the dispersion curves of SMPs.

The question of whether the group velocity will reverse in one-way systems can be answered by determining if the system's symmetry or chirality is broken. As demonstrated in Fig. 3(c), within the lower one-way region of the 'EYYE-r' arm, the SMPs can propagate only in the forward direction, regardless of which layer has a higher H_0 (ω_0). We consider two conditions, $[0.6, 0.4, r]$ and $[0.4, 0.6, r]$, where the propagation directions are the same. This is because the second case can be treated as the entire system of the first case revolving 180 degrees around the propagation direction, and thus the system's symmetry/chirality is conserved. In contrast, $[0.6, 0.4, s]$ and $[0.4, 0.6, s]$ have opposite propagation directions because they cannot be obtained by simply rotating each other, and thus the system's symmetry/chirality is broken when changing ω_0 accordingly.

To achieve a relatively broad one-way band, it is necessary that ω_0^c in arm 'B' is small enough, as shown in Fig. 3(d). Thus, we select $\omega_0^c = 0.1\omega_m$ (marked by the red dashed line in Fig. 3(d)), and Fig. 3(e) depicts the corresponding AFs and one-way regions as functions of ω_0^b . For this case, $\omega_{sp}^{(-2)} \approx 0.048\omega_m$ and $\omega_{sp}^{(+2)} = \omega_m + \omega_b$. With $\omega_0^a = 0.6\omega_m$ and $\omega_0^c = 0.1\omega_m$ fixed, the only remaining unknown parameter in the Y-shaped structure is ω_0^b . Ideally, we aim for the whole one-way region with $v_g < 0$ in arm 'B' to be the working band of the LGs. Based on our calculations, we can achieve this goal if $0 < \omega_0^b < 0.31\omega_m$. However, in most cases, to ensure that the entire one-way region of arm 'B' is the working band of LGs, we need to ensure that $\omega_{sp}^{(-2)}$ (the blue line in Fig. 3(d)) and $\omega_{sp}^{(+2)}$ (the green line in Fig. 3(d)) are both inside the one-way regions with $v_g > 0$ in arms 'A' and 'C'. To accomplish this, we construct the following equations:

$$\begin{cases} y_1 = (\omega_{sp_B}^{(-2)} - \omega_{sp_A}^{(-2)})(\omega_{sp_B}^{(-2)} - \omega_{sp_A}^{(+2)}) \\ y_2 = (\omega_{sp_B}^{(-2)} - \omega_{sp_C}^{(-2)})(\omega_{sp_B}^{(-2)} - \omega_{sp_C}^{(+2)}) \\ y_3 = (\omega_{sp_B}^{(+2)} - \omega_{sp_A}^{(-2)})(\omega_{sp_B}^{(-2)} - \omega_{sp_A}^{(+2)}) \\ y_4 = (\omega_{sp_B}^{(+2)} - \omega_{sp_C}^{(-2)})(\omega_{sp_B}^{(-2)} - \omega_{sp_C}^{(+2)}) \end{cases} \quad (8)$$

where 'A/B/C' represent arm 'A'/'B'/'C'. In this context, it is worth noting that arms 'A' and 'C' belong to the 'EYYE-r' type, while arm 'B' belongs to the 'EYYE-s' type. The AFs are represented by $\omega_{sp}^{(+)}$ and $\omega_{sp}^{(-)}$, which are given by Eqs. (3), (4), (6), and (7). Equation (8) determines whether $\omega_{sp}^{(-2)}$ in arm 'B' lies within the one-way region of arm 'A', which occurs for $y_1 < 0$. If $y_1 < 0$, $y_2 < 0$, $y_3 < 0$, and $y_4 < 0$ at the same time, it means that the entire one-way region with $v_g < 0$ in arm 'B' lies within the one-way regions of both arms 'A' and 'C'. Figure 3(f) represents the functions of y_1 ((f-1)), y_2 ((f-2)), y_3 ((f-3)), and y_4 ((f-4)) based on ω_0^b and ω_0^c when $\omega_0^a = 0.6\omega_m$. We observe that y_1 , y_2 , and y_4 are always negative, while y_3 can be positive for relatively large ω_0^b and small ω_0^c . Therefore, we set $\omega_0^a = 0.6\omega_m$ and $\omega_0^c = 0.1\omega_m$, and keep ω_0^b relatively small, such as $\omega_0^b = 0.1\omega_m$ (marked by red balls in Fig. 3(f)), to ensure that the entire one-way region in arm 'B' corresponds to the working band of LGs.

Note that the bandwidth of the one-way region in the three arms can be effectively controlled by adjusting the external magnetic field (ω_0). Based on our analysis and Fig. 3, it is possible to achieve a relatively broad one-way band, which is in general a desirable feature. Additionally, we can narrow the one-way band in specific arms to program the route for signal transmission. For example, when we set $\omega_0^a = \omega_0^c = 0.6\omega_m$ and $\omega_0^b = 0.1\omega_m$, the one-way channel between arms 'B' and 'C' will be closed, as indicated by the double-sided arrow in Fig. 3(d). In this scenario, only the wave input from arm 'A' can be transmitted to arm 'C'. By employing a similar strategy,

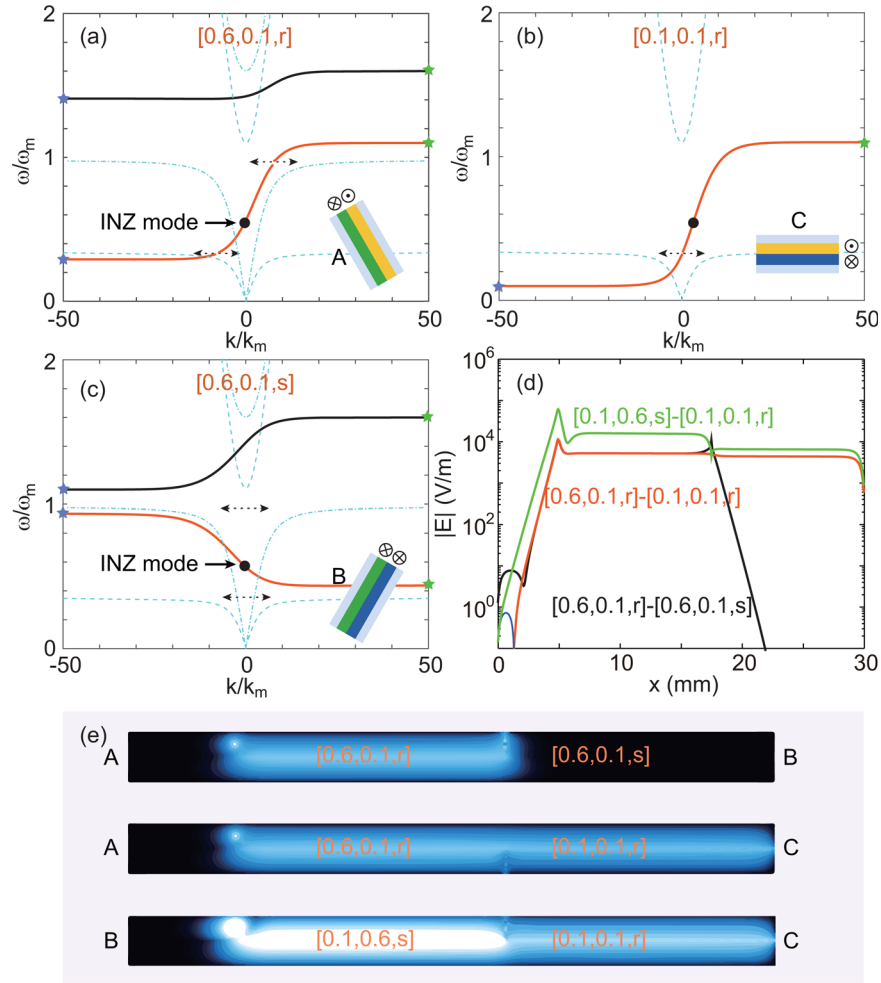


Fig. 4. (a-c) Dispersion diagrams of three arms with optimized parameters, $\omega_0^a = 0.6\omega_m$, $\omega_0^b = 0.1\omega_m$ and $\omega_0^c = 0.1\omega_m$. (d,e) The simulated electric field distribution obtained from coupling simulations containing two arms, with each arm being either the 'EYYE-r' type or 'EYYE-s' type.

we can design the transmission route according to our desires, providing greater control over signal propagation.

Figures 4(a)–4(c) present dispersion curves for the scenario where $\omega_0^a = 0.6\omega_m$ and $\omega_0^b = \omega_0^c = 0.1\omega_m$. Similar to Fig. 2, there is a one-way region with $v_g > 0$ in arm 'A' (depicted as a red-line region in Fig. 4(a)) and arm 'C' (depicted as a red-line region in Fig. 4(b)). Additionally, there is a one-way region with $v_g < 0$ in arm 'B' (depicted as a red-line region in Fig. 4(c)). Moreover, the backward one-way region is much larger than that illustrated in Fig. 2(c). Consequently, the working band of the LGs in this situation should be significantly broader. Figures 4(d), 4(e) show the coupling effect between arms when $f = 0.8f_m$, which falls within the one-way regions of interest. Consequently, two one-way channels ('A-C' and 'B-C') are established, while the EM signal cannot propagate from arm 'A' to arm 'B'. It should be noted that the first part ($[0.1, 0.6, s]$) of the 'B-C' channel differs from that of the 'A-B' channel ($[0.6, 0.1, s]$) due to the geometrical relationship between the arms. As per symmetry, the SMPs in the $[0.1, 0.6, s]$ and

[0.6, 0.1, s] structures must have opposite propagation directions. In the simulations, the EM signal can transfer efficiently in the one-way channels, while the forward transferring signal halts at the interface of arm 'A' and arm 'B'.

It is worth noting that despite not herein utilizing a physical theory or model to analyze the impact of nonlocal effects on our proposed structure, we believe that they can still achieve near-immunity to these effects through the tunability of the structure. Indeed, by appropriately adjusting the external magnetic fields, we can make the wavenumbers of the majority of the operating modes less than $10k_m$, as demonstrated in Figs. 4(a-c). Moreover, by selecting a specific frequency (indicated by the black dots in Figs. 4(a) and 4(c)), we can achieve index-near-zero (INZ) modes with wavenumbers close to zero. In this scenario, the nonlocal effect on the system should be negligible, particularly in the presence of realistic dissipative losses [17,40]. Therefore, the Y-shaped structure proposed in this paper has the potential to realize immunity to nonlocal effects. However, further careful theoretical studies are still required to fully validate this, and we plan to conduct these studies in our forthcoming research.

4. Realization of basic logic gates

The Y-shaped structure is designed to function within the one-way region and can serve as basic logic gates, including but not limited to OR, AND, and NOT gates, as illustrated in Fig. 5. During logical operations, arms 'A' and 'B' are the input ports, while arm 'C' is the output port. The structure operates as a natural OR gate, where any input one-way EM signal must propagate to the output port. If we assume that the presence of the EM signal is logic '1' and its absence is logic '0' (positive logic), then the Y-shaped structure functions as a versatile OR gate that can be externally adjusted by magnetic fields. However, the AND and NOT gates use negative logic, where the presence of the EM signal is recognized as logic '0.' In the AND operation, any input EM signal is treated as logic '0,' resulting in the output EM signal also being logic '0.' On the other hand, the NOT operation employs negative logic in either the input or output port, with positive logic used in the remaining port. Figure 5(b) shows simulations of the Y-shaped structure with $f = 0.8f_m$ when the EM signal is excited in only one of the input ports. Imperfections on the YIG-YIG interface were introduced as air holes with $r = 0.5$ mm to demonstrate the robustness of our proposed LGs. As a result, the LGs exhibits high CR of over 200 dB (infinity in theory). Figure 5(c) presents the corresponding truth tables for the OR, AND, and NOT operations.

As mentioned earlier in Fig. 3, we can control the LGs based on the Y-shaped structure with external magnetic fields (H_0). By changing the direction(s) of H_0 , we can reverse the propagation direction of the one-way SMPs, as illustrated by the lower one-way regions in Fig. 3(c,d), such as [0.6, 0.4, r] transitioning to [0.6, 0.4, s]. Another notable case is that changing the direction(s) of H_0 can cause the previous one-way region to close, such as [0.6, 0.6, r] shifting towards [0.6, 0.6, s]. In this scenario, there are no one-way regions, and the entire band becomes a band gap, preventing the propagation of EM signals. Additionally, altering the value of H_0 , either by increasing or decreasing it, can significantly affect the logic operations. Therefore, the operating band of our proposed LGs can be easily tuned by changing the external magnetic fields.

Besides, we suggest an innovative approach to tune LGs, as demonstrated in Fig. 6. This method achieves three modes of LGs by switching H_0 or ω_0 , namely the "work," "stop," and "skip" modes. As initial magnetic-field parameters, we set $\omega_0^a = 0.6\omega_m$, $\omega_0^b = -0.1\omega_m$, and $\omega_0^c = 0.1\omega_m$. It is noteworthy that the '-' sign indicates the external magnetic field is in the +z direction. Upon exchanging ω_0^b (H_0^b) and ω_0^c (H_0^c), it is easy to calculate that SMPs with $f = 0.8f_m$ have opposite propagation directions in the original arms. Similarly, exchanging ω_0^c (H_0^c) and ω_0^a (H_0^a) reverses the propagation direction of SMPs in arm 'B,' whereas the direction remains unchanged in other arms. Moreover, exchanging ω_0^a (H_0^a) and ω_0^b (H_0^b) reverses the propagation direction of SMPs in arm 'A,' whereas the direction remains unaltered in other arms. The simulation of the three modes of LG is illustrated in Figs. 6(b) and 6(c). The original mode is

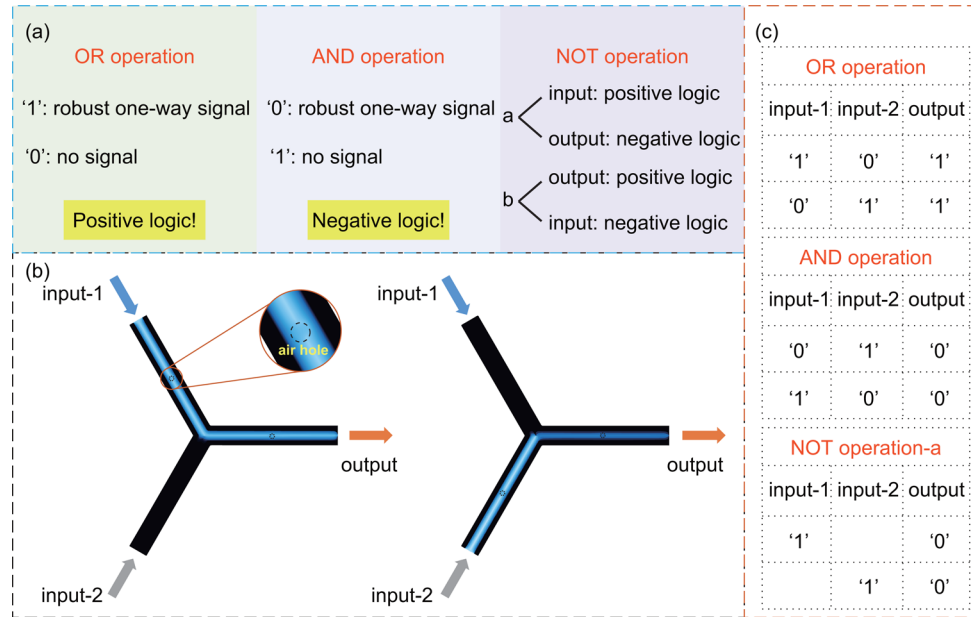


Fig. 5. (a) Theory of all-optical logic operation using the positive and/or negative logic. (b) Numerical simulations in the Y-shaped module as $f = 0.8f_m$, and air holes with $r = 0.5 \text{ mm}$ were set on the YIG-YIG interfaces to verify the robustness of logic operations. (c) The truth tables of the OR, AND, and NOT operations.

designated the "work" mode since it can work as LGs. Two methods can achieve the "stop" mode with EM signals being halted, and one method can accomplish the "skip" mode with EM signals skipping the present calculation. Therefore, our proposed Y-shaped LGs offers rich manipulation possibilities and is promising for programmable optical communication/devices. In contrast, traditional all-optical LGs typically operate at fixed frequencies and can be challenging to tune.

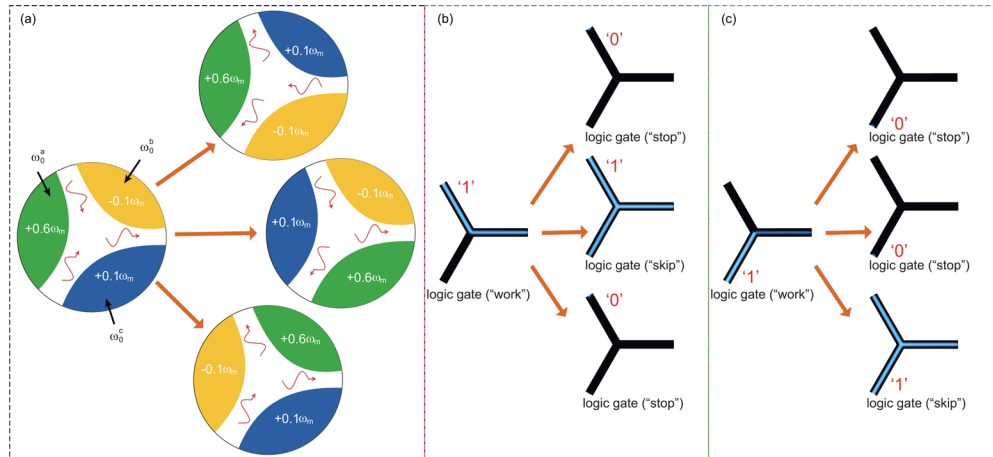


Fig. 6. (a) Switch theory for tunable LGs. The red arrow refers to the EM wave path that allows passage. (b,c) Simulation for tunable LGs by switching the external magnetic fields for input signals (b) [1', 0'] and (c) [0', 1'] using positive logic. Input '1' could be alternatively programmed to '0' ("stop" mode) or '1' ("skip" or "work" mode.)

5. Conclusion

In summary, we have proposed a Y-shaped structure made of YIG layers with distinct magnetizations. The arms of the structure were categorized into two types: The 'EYYE-r' type with opposing magnetization directions, and the 'EYYE-s' type with identical magnetization directions. Our theoretical analysis of the 'EYYE-r' and 'EYYE-s' arms led to the construction of two one-way channels capable of supporting topologically protected one-way SMPs. Furthermore, the implementation of basic logic gates, such as OR, AND, and NOT gates, is achieved through these broadband and topological one-way SMPs, resulting in highly robust (resistant to backscattering and imperfections) and precise (theoretically infinite contrast ratio) LGs. In addition, we explored the tunability of these LGs. By adjusting external magnetic fields, the one-way region can be easily modulated, either broadened or narrowed, the propagation directions of the SMPs within the region can be completely reversed, and the region can be closed. Given the intriguing tunability of the operating band of the Y-shaped LGs. In addition, we proposed a practical application for the structure/LGs: By switching external magnetic fields, three switchable modes ("work", "skip", and "stop") can be achieved. Our proposed magnetized YIG-based LGs have the potential to enable tunable all-optical logic operations and may lead to high-efficiency and programmable all-optical microwave communication. The discovery or engineering of tunable one-way modes could pave the way for achieving all-optical LGs in higher frequency regimes, such as terahertz or near-infrared, using similar methods as proposed in this paper. These advances could have significant implications for future optical communication systems and signal processing applications.

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